

No. 1820C

LA1265

FM/AM TUNER OF ELECTRONIC TUNING TYPE

Functions

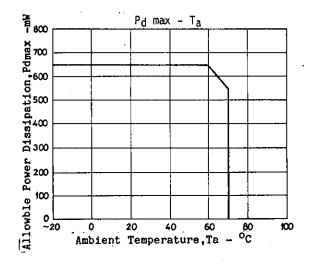
FM: IF amp, quadrature detector, AF preamp, signal meter, tuning indicator drive output (common with stop signal, muting drive output)

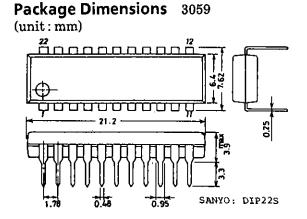
AM: RF amp, MIX, OSC (with ALC), IF amp, detector, AGC, signal meter, tuning indicator drive output (common with stop signal).

Features

- . Minimum number of external parts required.
- . Excellent S/N
- . Local OSC with ALC
- . Local OSC buffer
- . Tuning indicator pin (common with narrow-band stop signal and muting drive output)
- . Variable stop sensitivity (variable separately for FM, AM)
- . Low whistle
- . Signal meter pin.

Maximum Ratings at Ta=25°C, Se	e Test C	ircuit.		unit
Maximum Supply Voltage	V_{CC} max	Pins 7,8,17	16	V
Flow-in Current	18	Pin 8	20	mA
Flow-out Current	120	Pin 20	1	. mA
	I ₂₂ Pdmax	Pin 22	2	mA
Allowable Power Dissipation	Pdmax	Ta=≦60°C	650	$\mathbf{m} \mathbf{W}$
Operating Temperature	Topr		-20 to +70	°C
Storage Temperature	Tstg	•	-40 to +125	°C
Operating Conditions at Ta=25°	c c			unit
Recommended Supply Voltage	v_{cc}		8.5	V
Operating Voltage Range	V _{CC} op		6 to 14	V

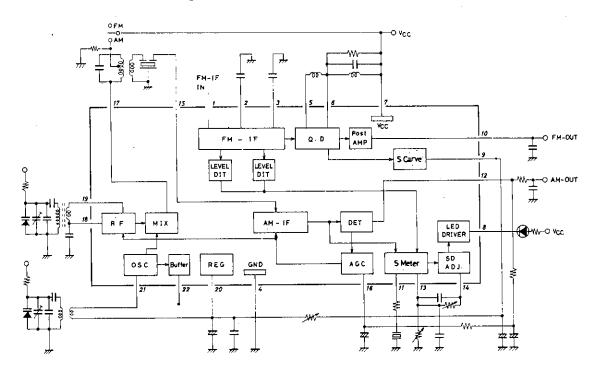




Operating Characteristics a	t Ta=25	°C,V _{CC} =8.5V,See Test Circu	it.			
[AM: f=1MHz]			min	typ	max	unit
Quiescent Current	Icco	No input		18	26	$\mathbf{m}\mathbf{A}$
Detection Output	Vo(1)	Vi=20dBu, 400Hz, 30% mod.	30	50	90	mV
		Vi=80dBu,400Hz,30% mod.	110	160	220	$\mathbf{m}V$
S/N	S/N(1)	Vi=20dBu	16	20		dΒ
	S/N(2)	Vi=80dBu	49	54		đΒ
Total Harmonic Distortion	THD(1)	V1=80dBu,400Hz,30% mod.		0.3	1.0	%
		Vi=107dBu,400Hz,30% mod.		0.5	2.0	%
Signal Meter Output	V _{SM(1)}	No input	0	0	0.2	v
	V <u>SM(2)</u>	<u>.</u>	2.4	2.8		V
LED Lighting Sensitivity	LED on	I _{LED} =1mA	15	24	33	dBu
Local OSC Buffer Output	Vosc	fose=1.45MHz	220	275	330	mV
		OSC		-		
[FM: $f=10.7MHz$]			min	typ	max	unit
Quiescent Current	Icco	No input		20	28	$\mathbf{m}\mathbf{A}$
Input Limiting Sensitivit	y -3dBL	.S 3dBdown, 400Hz, 100% mod.		31	37	dBu
Demodulation Output	Vo	Vi=100dBu, 400Hz, 100% mod.	240	330	460	mV
S/N	S/N	Vi=100dBu	78	84		dВ
Total Harmonic Distortion	THD	Vi=100%dBu,400Hz,100% mod		0.03	0.3	%
Signal Meter Output		No input	0	Ō	0.2	v
	V _{SM} (2)	Vi=100dBu	1.5	2.7	3.1	V
LED Lighting Sensitivity	LED-on	I _{LED} =1mA	35	-	65	dBu
LED Lighting Bandwidth	LED-BW	VI=100dB,I _{LED} =1mA	90		160	kHz
AM Rejection	AMR V:	i=100dBu,FM=400Hz 100% mod	-	60		dB
-		M=1kHz 30% mod.	_			

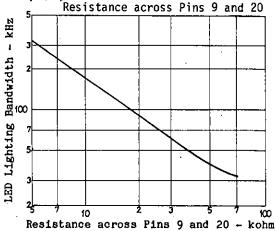
Note: Be fully careful of dielectric breakdown.

Equivalent Circuit Block Diagram



How to use the LA1265

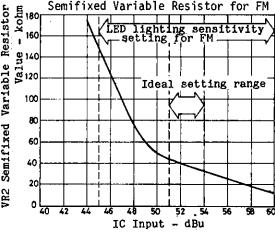
- 1. LED lighting, muting drive output, stop signal (S.D) LED Lighting Bandwidth -
- ① For LED lighting, muting drive output, stop signal, the output at pin 8 is used.
- ② The voltage on pin 8, when tuned, turns from "H" to "L". (Active-Low)
- 3 Signal bandwidth at pin 8
 - . For AM, the bandwidth depends on the CF (BFU450CN) at pin 11. If a capacitor is connected in place of the CF, the bandwidth will get wider.
 - For FM, the bandwidth depends on the resistance across pins 9 and 20. If the resistance is increased, the bandwidth will get narrower. R=15kohms makes the bandwidth approximately 120kHz.



- 4 Sensitivity adjustment of LED, muting, stop signal
 - . For FM, the semifixed variable resistor across pin 13 and GND is used.
 - For AM, the semifixed variable resistor across pins 13 and 14 is used. Be sure to start adjustment for FM, and then make adjustment for AM. For the stop signal sensitivity and FM stop signal bandwidth, the variations should be considered and it is recommended to use the semifixed variable resistor for adjustment.
- (5) LED lighting sensitivity setting for AM

 For the LED lighting sensitivity setting for AM, it is desirable that the IC input be 30dBu (antenna input: approximately 50dB/m). In this case, the value of VR1 is 30kohms.

 Semifixed Variable Resistor for FM
- For the LED lighting sensitivity setting for FM for the LED lighting sensitivity setting for FM, the IC input may be 45dBu to 8 140 60dBu. With the variations in the front end considered, it is ideal that the IC input in a standard receiving set be 51dBu to 54dBu. The lower value of VR2 for the LED lighting sensitivity setting is as illustrated right. Since the variations in the front end cause the IC input setting sensitivity to vary, it is recommended to use a value of VR2 at an input voltage lower than a standard setting by 6dB or greater. For example, if



- IC inut 53dBu is taken as a standard, use VR2=100kohms at IC input 47dBu.
- 2. AM/FM changeover
- Two selections are available for changeover as shown below: (A) pin 17-used method and (B) pin 18-used method.
- 3) For (A), a resistance of 68kohms at the IFT cold terminal, which is used to prevent the changeover circuit from malfunctioning, must be connected.

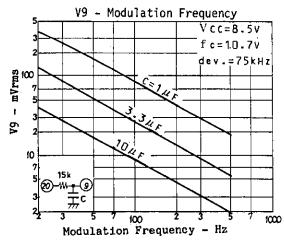
Unit (resistance: Ω)

AM/FM changeover

(A) Pin 17-used method for

(B) Pin 18-used method for AM/FM changeover

- 3. Local OSC buffer output
- ① When local OSC buffer output waveform is saw-toothed at the SW mode, connect a resistance of 1.2kohms or thereabouts across pin 22 and GND.
- 4. AM input pin
- ① It is desirable that the AM input pin (pin 19) be L-coupled to pin 18.
- ② Inputting to pin 19 can be done by DC-cutting with a capacitor. However, an unbalance in the RF amplifier (differential amplifier) causes gain drop and whistle worsening.
- 5. Capacitance across pin 9 and GND A large capacitance across pin 9 and GND may cause a misstop at an adjacent channel when the channel select speed is made faster at the automatic channel select mode. In this case, decrease the capacitance across pin 9 and GND. However, if too decreased, the LED will flutter at & low modulation frequencies at the time of detuning. Therefore, it is recommended to fix the capacitance across pin 9 and GND to be 3.3uF to 10uF. The relation between modulation frequency and demodulation output voltage on pin 9 with the capacitance across pin 9 and GND as a parameter is shown right.



- 6. If the coupling coefficient of the local OSC coil is small and an antiresonance point of approximately 100MHz is present or the stray capacitance across pins 22 and 21 is large, a parasitic oscillation of approximately 100MHz may occur in the buffer output (pin 22). In this case, connect a capacitance of approximately 30pF across pin 22 and GND.
- 7. AM OSC coil

Generally speaking, the following should be noted. Avoid winding with loose coupling between primary side and secondary side (especially SW1, SW2). To put it concretely, the pot core type is better than the screw core type which is loose in coupling. This prevents the local OSC frequency from turning third resonance frequency related to the coupling coefficient. v_{DD}

8. Resistance across pin 8 and V_{DD}

If pin 8 is used for the stop signal (SD) only, without using LED, it is recommended to fix resistance R_L across pin 8 and sp GND to be 51kphms to 100kphms.

GND to be 51kohms to 100kohms.

9. To prevent whistle from worsening, make the pattern of AM output pin 12 as short as possible.

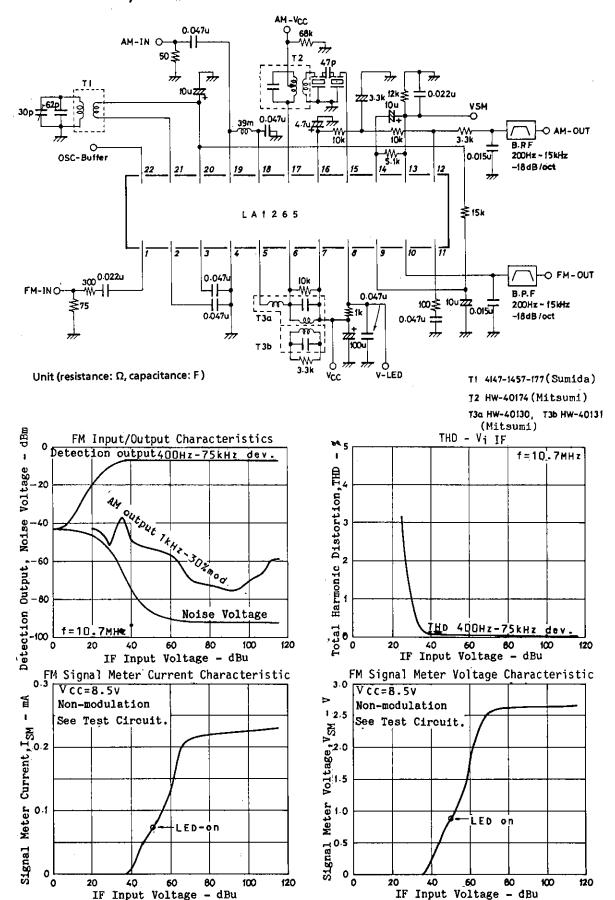
Input/Output Admittance

FM					
-	Parameter	Frequency	-	Admittance	unit
I F	y i 1	10.7MHz	ri	330	Ω
	1		Сi	20	рF

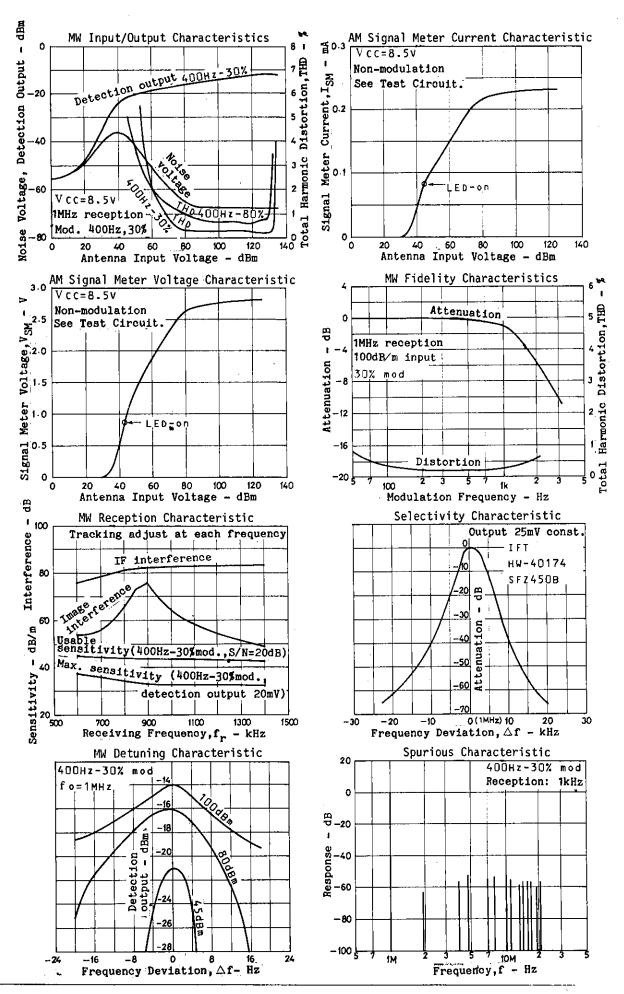
- Parameter	Frequency	-	Admittance			
			AGC-off(V16=1.4V)	AGC-on(V16=2.5V)		
RF	yi19 1MHz	19 1MHz	ri	15	16	kΩ
		Сi	4	4	pF	
MIX	y o 17 500kHz	Гo	-		kΩ	
			Co	3	3	рF
IF y 115 500kHz	ri	2	2	kΩ		
	· .	C o	10	8	рF	

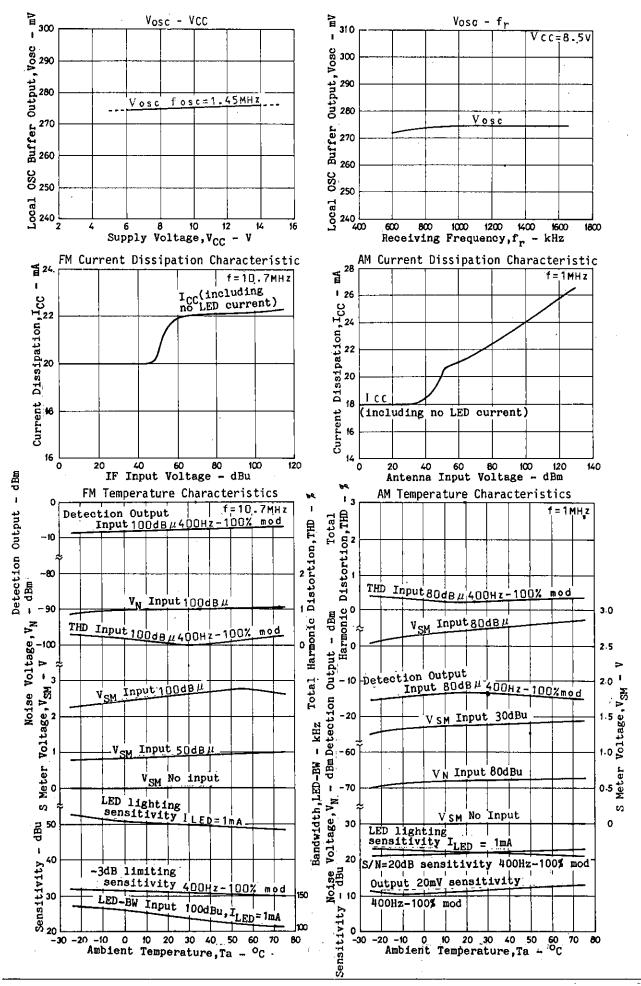
A F LA1265

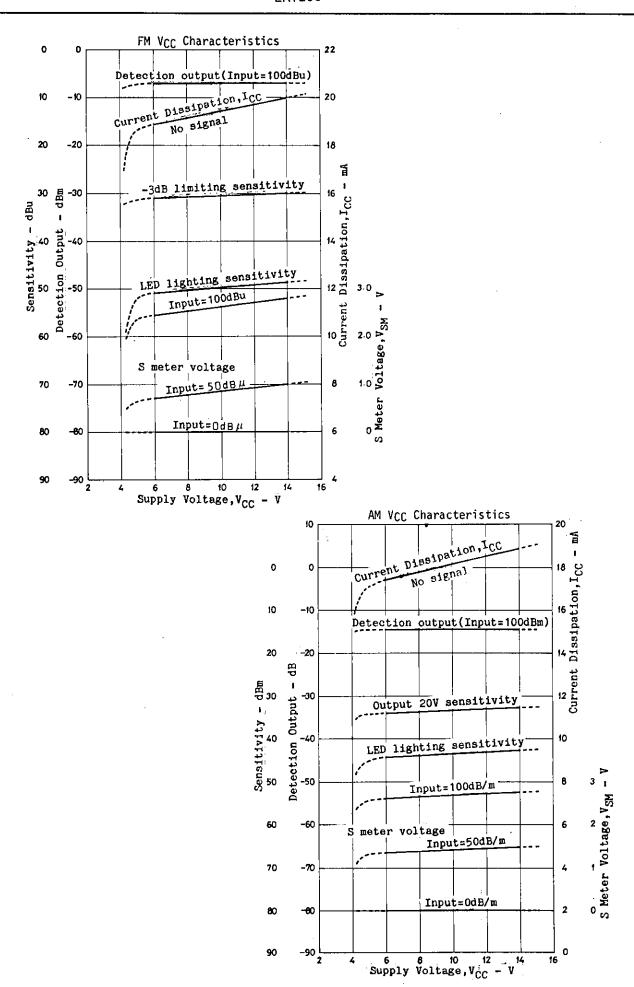
Test Circuit : FM, AM-MW



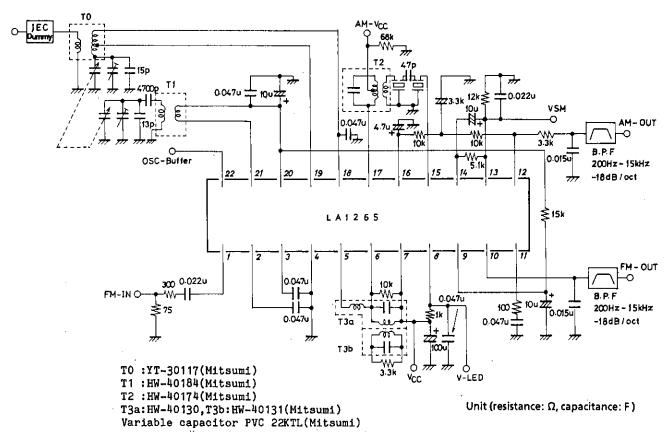
IF Input Voltage - dBu

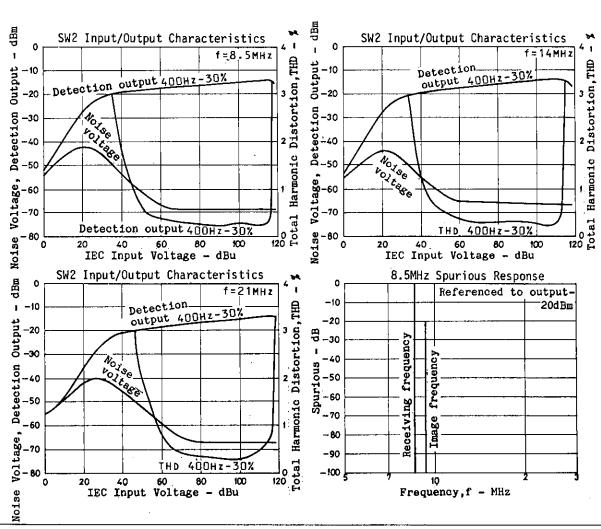


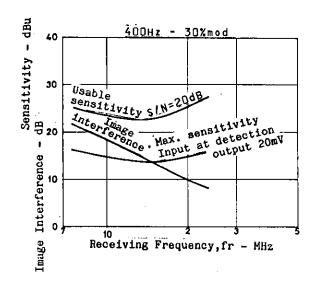


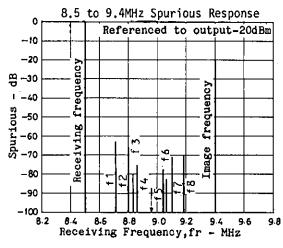


Test Circuit : SW2









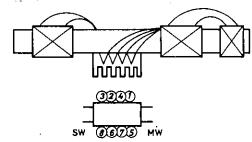
f1: $8.724\text{MHz} \rightarrow 2\text{fosc-}2\text{f1}=455\text{kHz}$ f2: $8.799\text{MHz} \rightarrow 3\text{fosc-}3\text{f2}=455\text{kHz}$ f3: $8.836\text{MHz} \rightarrow 4\text{fosc-}4\text{f3}=455\text{kHz}$ f4: $8.859\text{MHz} \rightarrow 5\text{fosc-}5\text{f4}=455\text{kHz}$ f5: $9.038\text{MHz} \rightarrow 5\text{f5-}5\text{fosc-}455\text{kHz}$ f6: $9.061\text{MHz} \rightarrow 4\text{f6-}4\text{fosc-}455\text{kHz}$ f7: $9.098\text{MHz} \rightarrow 3\text{f7-}3\text{fosc-}455\text{kHz}$ f8: $9.173\text{MHz} \rightarrow 2\text{f8-}2\text{fosc-}455\text{kHz}$

Coil Specifications

MW antenna

Bar antenna (for PVC22KTL)

-TN-10896(Mitsumi)



①-② 22T+49T,③-④ 10T

Tight solenoid direct winding

⑤-⑥ 17T 0.56 space winding

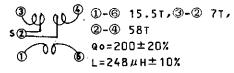
7-8 4T tight solenoid winding

①-② L=260 μ H,Qo=330(\geq 200)

⑤-⑥ L=15μH,Qo=250(≥150)

Loop antenna (for SVC321)
-LA300(Korin Giken)
Loop antenna matching coil

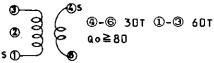
·KL-412



MW OSC

-4147-1457-177(Sumida)

For PVC22KTL



-K0-387(Korin Giken) For SVC321

① ① -③ 48T,④-⑥ 20T ② ② ② Qo=150±20% S ① —

AM-IFT

: Matching coil for SFU450B (1-element type)



-HW-40173 (Mitsumi)

①-② 821,◎-② 701,

Qo=110±20%,f=450kHz

Internal 180pF

2150-2162-197(Sumida)

①-② 103T,③-② 71T,

⊕-© 8T

Q≧80,f=450kHz

Internal 180pF

Matching coil for SFZ450B (2-element type)



·HW-40174(Mitsumi) ①-② 58T,③-② 94T,

④-⑥ 101

Qo=80±20%,f=450kHz

Internal 180pF

-2150-2061-049 (Sumida)

①-② 54T, ③-② 12OT,

4-6 12T

Qo≧40

Internal 180pF

FM single tuning detection coil



'HW-40122 (Mitsumi)

③-④ 84.5T,③-① 19T Qo=35±20%,f=10.7MHz

Internal 82pF±10%

Damping resistance

-2231-016 (Sumida)

3-4 73.5T, 3-1 19T

Qo=30±20%,f=10.7MHz

Internal 82pF±10%

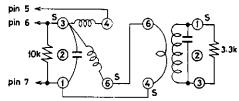
Damping resistance

FM double tuning detection coil



HW-40131

2231-097



·HW-40130 (Mitsumi) ·HW-40131 (Mitsumi)

3-4 86.5T

Ф-© 1т

③-⑥ 13.5T

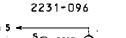
①-③ 19T

Qo#50±20%

Qo=35±20%

Internal 100pF=10%

Internal 100pF±10%



<u>\$</u>(3) 10×≦②

-2231-096 (Sumida)

-2231-097(Sumida)

3-9 73.5т

@-© 2T

3-6 23.5T

①-③ 21T

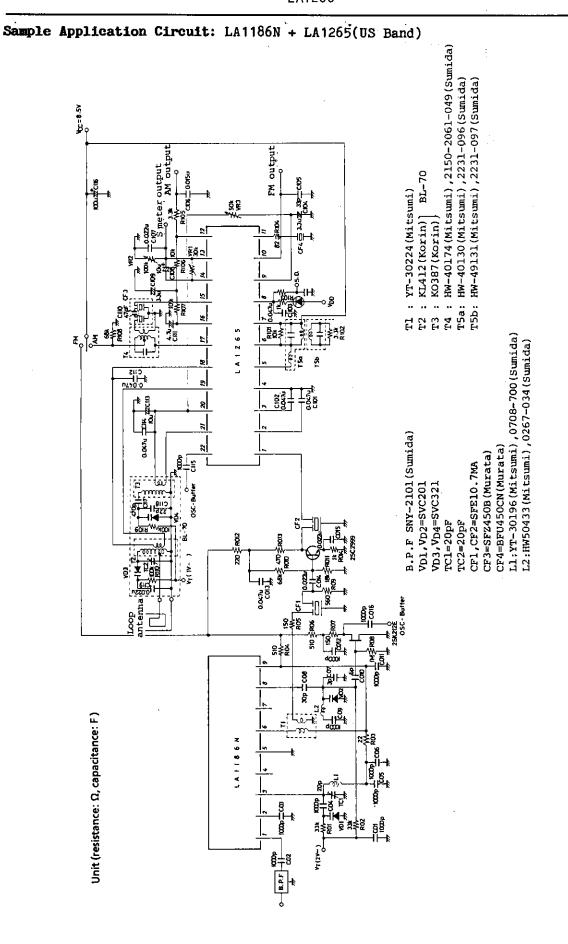
Qo=50±20%

Qo=47±20%

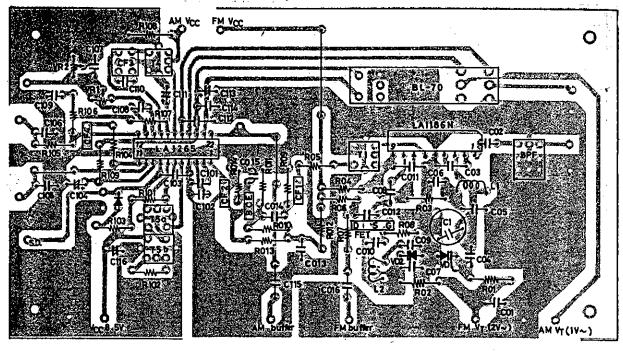
Internal 62pF±10%

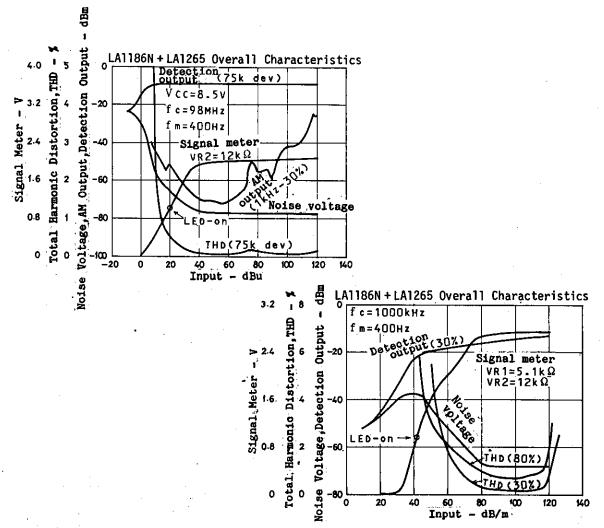
Internal 82pF±10%

Unit (resistance: Ω)



Example of Printed Pattern (Cu-foiled area)





- No products described or contained herein are intended for use in surgical implants, life-support systems, aerospace equipment, nuclear power control systems, vehicles, disaster/crime-prevention equipment and the like, the failure of which may directly or indirectly cause injury, death or property loss.
- Anyone purchasing any products described or contained herein for an above-mentioned use shall:
 - Accept full responsibility and indemnify and defend SANYO ELECTRIC CO., LTD., its affiliates, subsidiaries and distributors and all their officers and employees, jointly and severally, against any and all claims and litigation and all damages, cost and expenses associated with such use:
 - ② Not impose any responsibility for any fault or negligence which may be cited in any such claim or litigation on SANYO ELECTRIC CO., LTD., its affiliates, subsidiaries and distributors or any of their officers and employees jointly or severally.
- Information (including circuit diagrams and circuit parameters) herein is for example only; it is not guaranteed for volume production, SANYO believes information herein is accurate and reliable, but no guarantees are made or implied regarding its use or any infringements of intellectual property rights or other rights of third parties.